# Mathematics of Image and Data Analysis Math 5467

# Lecture 16: The Sampling Theorem and Cosine Transform

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## Last time

- Mult-dimensional DFT
- Image denoising

## Today

- The sampling theorem
- Discrete Cosine Transform

### The Sampling Theorem

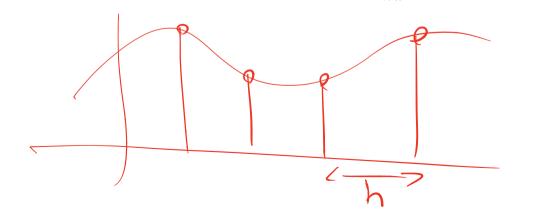
If a signal  $u : \mathbb{R} \to \mathbb{R}$  contains no frequencies greater than  $\sigma_{max}$ , then u can be perfectly reconstructed from evenly spaced samples provided the sampling frequency is greater than the Nyquist rate  $2\sigma_{max}$  and we have the Sinc Interpolation formula

 $\operatorname{sinc}(x) = \frac{\sin(\pi x)}{1 - \frac{1}{2}}$ 

1 > 2 0 max

(1) 
$$u(t) = \sum_{j=-\infty}^{\infty} u(jh) \operatorname{sinc}\left(\frac{t-jh}{h}\right),$$

where h is the sampling period and



# The Sampling Theorem

If a signal  $u : \mathbb{R} \to \mathbb{R}$  contains no frequencies greater than  $\sigma_{max}$ , then u can be perfectly reconstructed from evenly spaced samples provided the sampling frequency is greater than the Nyquist rate  $2\sigma_{max}$  and we have the Sinc Interpolation formula

 $\int_{-\infty}^{\infty} \frac{2\pi i t k}{i f |k|^2} dt = 0$ 

(2) 
$$u(t) = \sum_{j=-\infty}^{\infty} u(jh) \operatorname{sinc}\left(\frac{t-jh}{h}\right),$$

where h is the sampling period and

$$\operatorname{sinc}(x) = \frac{\sin(\pi x)}{\pi x}.$$

- The sampling frequency is  $\frac{1}{h}$ , so the Nyquist rate condition for the Sampling Theorem is that  $\frac{1}{h} > 2\sigma_{max}$ , or  $h < \frac{1}{2\sigma_{max}}$ .
- At sampling intervals  $h > \frac{1}{2\sigma_{max}}$ , high frequencies are aliased to lower frequencies, creating distortion.
- CD quality audio samples at a rate of 44.1 kHz, which was chosen to capture frequencies up to 22.05 kHz, higher than most humans can hear.

#### The Sampling Theorem (periodic version)

Let  $u : \mathbb{R} \to \mathbb{R}$  be periodic with period 1, and assume u has no frequency larger than  $\sigma_{max}$ , where  $\sigma_{max}$  is a positive integer. This means that the signal u has the Fourier Series representation

(3) 
$$C_{\mathbf{k}} \in \mathcal{C}$$
  $u(t) = \sum_{k=-\sigma_{max}}^{\sigma_{max}} c_k e^{2\pi i k t}$ . Fourier Series

**Theorem 1.** Suppose that u is given by (3) and let h = 1/n for  $n \in \mathbb{N}$  with  $n > 2\sigma_{max}$ . Assume also that n is odd. Then u(t) can be reconstructed from its evenly spaced samples u(jh) and furthermore we have

(4) 
$$u(t) = \sum_{j=0}^{n-1} u(jh) S\left(\frac{t-jh}{h}\right),$$

where S(t) is given by

$$S(t) = \frac{\operatorname{sinc}(t)}{\operatorname{sinc}(ht)}.$$

#### Sinc kernel

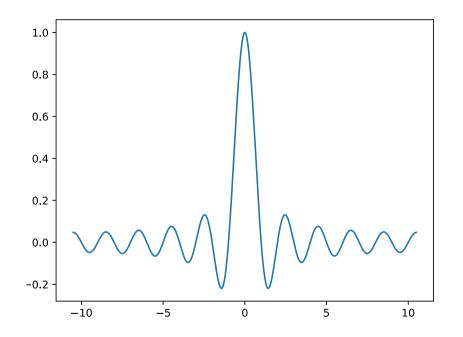


Figure 1: Depiction of the Sinc-like kernel  $S(t) = \operatorname{sinc}(t) / \operatorname{sinc}(ht)$  for n = 21 and h = 1/21. The kernel is periodic with period n = 21.

#### **Proof of Sampling Theorem**

So  $f \in L^2(\mathbb{Z}_n)$ 

Ult) = CKE  $, h = \frac{1}{n}$ K=-Smax

Sample U:  $f(j) = U(jh) = U(\frac{j}{n})$ 

f: 2-s R is n-periodic

Q: Can we recome Ck from f?

 $f(j) = U(\frac{j}{n}) = \sum_{k=-0}^{\infty} C_k e^{2\pi i k j/n}$ = Ckh, w=eth k=-Junx Try taking inner product : Up(k) = W Kl  $(f, u_e) = \sum_{i=0}^{n} f(i) \omega^{-ie}$ ,  $\overline{u_e}(\kappa) = \omega^{-\kappa e}$ -MEREN = Z CKW j(K-l) J=> K=-Jmx

 $w = e^{2\pi i/n} \quad (\varphi) = \sum_{k=-\sigma_{mx}}^{\sigma_{mx}} c_k \sum_{j=0}^{m} w^{j(k-\ell)} \neq n c_k$  $? \neq \leq 0, k \neq l$ (n, k = lRecall W=1 If k-l = pn (") then  $w^{j(k-e)} = w^{jpn} = 1$ . = N Z Ck S{kel mod n} k=- Omix

k-l=pn, pEZ. Kel mad n means It 25max CN C=> 25max Sh-1 then the sum above has only sue term Ce. In this case  $\langle f, ue \rangle = n C e_{n}, C e_{n} = \int_{n} \langle f, ue \rangle$ and  $u(t) = \int_{k=-\infty}^{\infty} \langle f, u_k \rangle e^{2\pi i k t}$ 

 $\int \sum_{k=-}^{\infty} \langle f, u_k \rangle e^{2\pi i k t}$   $\wedge k = -(\frac{\pi}{2})$ , Omex C h-1 Z Since  $(f, u_k) = 0$  for  $\mathcal{O}_{\max} \leq |k| \leq \frac{N-1}{7}$  $y = \frac{1}{2} \sum_{k=-(\frac{n-1}{2})}^{n-1} \frac{-2\pi i j k}{k} \frac{2\pi i k t}{k}$ u(jh)1=h /

= sinc(n(t-jh)) Sinc(+) ugh) Sinc(n(t-jh))  $l_{1(4)}$ Sinc (t-jh) 722 N=- $S\left(\frac{t-5h}{h}\right)$ S(t) = Sinc(t)Sinc (ht)

#### **Discrete Cosine Transform**

It is often useful in practical applications to avoid complex numbers and work with real-valued transformations. If  $f \in L^2(\mathbb{Z}_n)$  is *real-valued* then the Fourier Inversion Theorem yields

-2mijl/n 2mikl/n P x = EntilleY = LTTK// = (Cos X + i sin X)(Cos Y + i sin Y)= COSXCOSY - SINXSINY ] keep + i (sunx cosy + cosx siny). vanishes when summed abour, since fiks ER.

#### Even/odd extensions

Let  $f : \mathbb{Z}_n \to \mathbb{R}$ . We define the even extension  $f_e : \mathbb{Z}_{2(n-1)} \to \mathbb{R}$  by

(5) 
$$f_e(k) = \begin{cases} f(k), & \text{if } 0 \le k \le n-1, \\ f(2(n-1)-k), & \text{if } n \le k \le 2(n-1)-1. \end{cases}$$

The odd extension  $f_o: \mathbb{Z}_{2(n+1)} \to \mathbb{R}$  is defined by

(6) 
$$f_o(k) = \begin{cases} 0, & \text{if } k = 0\\ f(k-1), & \text{if } 1 \le k \le n, \\ 0, & \text{if } k = n+1\\ -f(2(n+1)-1-k), & \text{if } n+2 \le k \le 2(n+1)-1. \end{cases}$$

#### Even/odd extensions

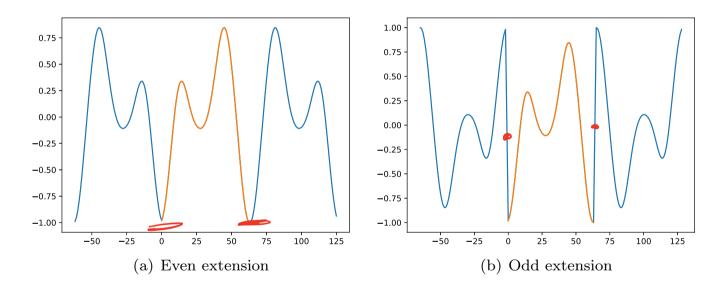


Figure 2: Example of the even and odd extensions of a signal on  $\mathbb{Z}_{64}$ 

#### **Discete Cosine Transform**

Recall

$$\begin{array}{l} (\clubsuit) & = \frac{1}{n} \sum_{\ell=0}^{n-1} \left( \sum_{j=0}^{n-1} f(j) \cos(2\pi j\ell/n) \right) \cos(2\pi k\ell/n) \\ & \quad + \frac{1}{n} \sum_{\ell=0}^{n-1} \left( \sum_{j=0}^{n-1} f(j) \sin(2\pi j\ell/n) \right) \sin(2\pi k\ell/n). \end{array}$$

We now apply the representation formula above to the even extension  $f_e$ , taking 2(n-1) in place of n, to obtain the Discrete Cosine Transform

$$f(k) = \frac{1}{2(n-1)} (A_0 + (-1)^k A_{n-1}) + \frac{1}{n-1} \sum_{\ell=1}^{n-2} A_\ell \cos\left(\frac{\pi k\ell}{n-1}\right),$$

 $\sim$ 

where

$$A_{\ell} = f(0) + (-1)^{\ell} f(n-1) + 2\sum_{k=1}^{n-2} f(k) \cos\left(\frac{\pi k\ell}{n-1}\right)$$

Apply (+) to te  $f_e(k) = \frac{1}{2(n-1)} \sum_{k=0}^{2(n-1)-i} A_e \cos\left(\frac{aTKe}{a(n-1)}\right)$ 

+ 1 2(n-1) ==> Be Sin (2(n-1))

 $A_{\ell} = \sum_{j=2}^{2} f_{e}(j) \cos\left(\frac{\pi j l}{p(n-i)}\right)$  $Be = \sum_{j=2}^{2(n-1)^{-1}} fe(j) \sin\left(\frac{a\pi j}{a(n-1)}\right)$ 

Claim? Be=0 (since fe reven) exercise! Let's compute AR.  $f_{e}(k) = \begin{cases} f(k), \quad 0 \le k \le n-1 \\ f(a(n-1)-k), \quad n \le k \le a(n-1)-1 \end{cases}$  $A_{\ell} = \sum_{j=0}^{n-1} f(j) \cos\left(\frac{\pi j \ell}{n-1}\right) + \sum_{j=n}^{2(n-1)-1} f(a(n-1)-j) \cos\left(\frac{\pi j \ell}{n-1}\right)$ 

j=n (=> k= n-2 k = 2(n-1) - jj=2(n-1)-1 (=> K=) j = 2(n-1) - k $C = \sum_{k=1}^{n-1} f(k) C_{95} \left( \frac{T(2(n-1)-k)R}{n-1} \right)$  $z \cos\left(2\pi \ell - \frac{\pi k \ell}{n-1}\right)$ = Cos(-Tkl) = Cos(Tkl) $C = \sum_{k=1}^{n-2} f(k) C_{2i} \left( \frac{T k \ell}{n-i} \right)$ 

 $A_{\ell} = f(o) + f(n-1) \cos(\pi \ell)$ +  $2\sum_{j=1}^{n-2} f(j) e_{2s} \left( \frac{\pi j l}{n-1} \right)$ Rest left as exercise (see class votes).

## **Discrete Sine Transform**

Using the odd extension we get the Discrete Sine Transform

(7) 
$$f(k) = \frac{1}{n+1} \sum_{\ell=0}^{n-1} B_{\ell} \sin\left(\frac{\pi(k+1)(\ell+1)}{n+1}\right),$$

where

$$B_{\ell} = 2\sum_{k=0}^{n-1} f(k) \sin\left(\frac{\pi(k+1)(\ell+1)}{n+1}\right),\,$$

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